



# Experience with RF systems at DELTA

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## DELTA parameters:

Beam energy: 550 MeV – 1.5 GeV  
Beam current: 130mA @ 1.5GeV  
Beam lifetime: 16h @ 130 mA  
Availability: 95 %  
Operational: 3000 h / year

## RF Group:

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# DELTA's RF systems

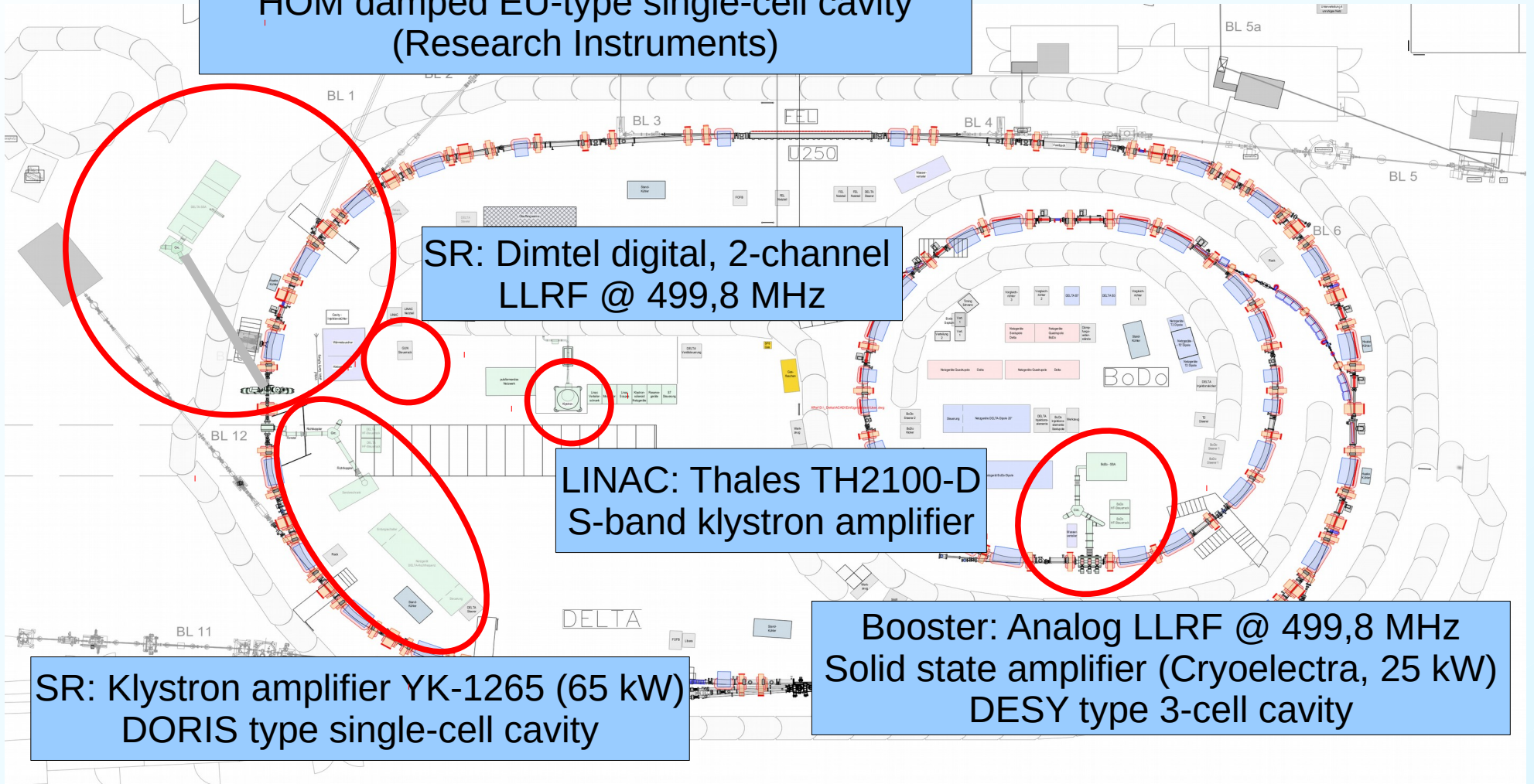
SR: Solid state amplifier (Cryoelectra, 75 kW)  
HOM damped EU-type single-cell cavity  
(Research Instruments)

SR: Dimtel digital, 2-channel  
LLRF @ 499,8 MHz

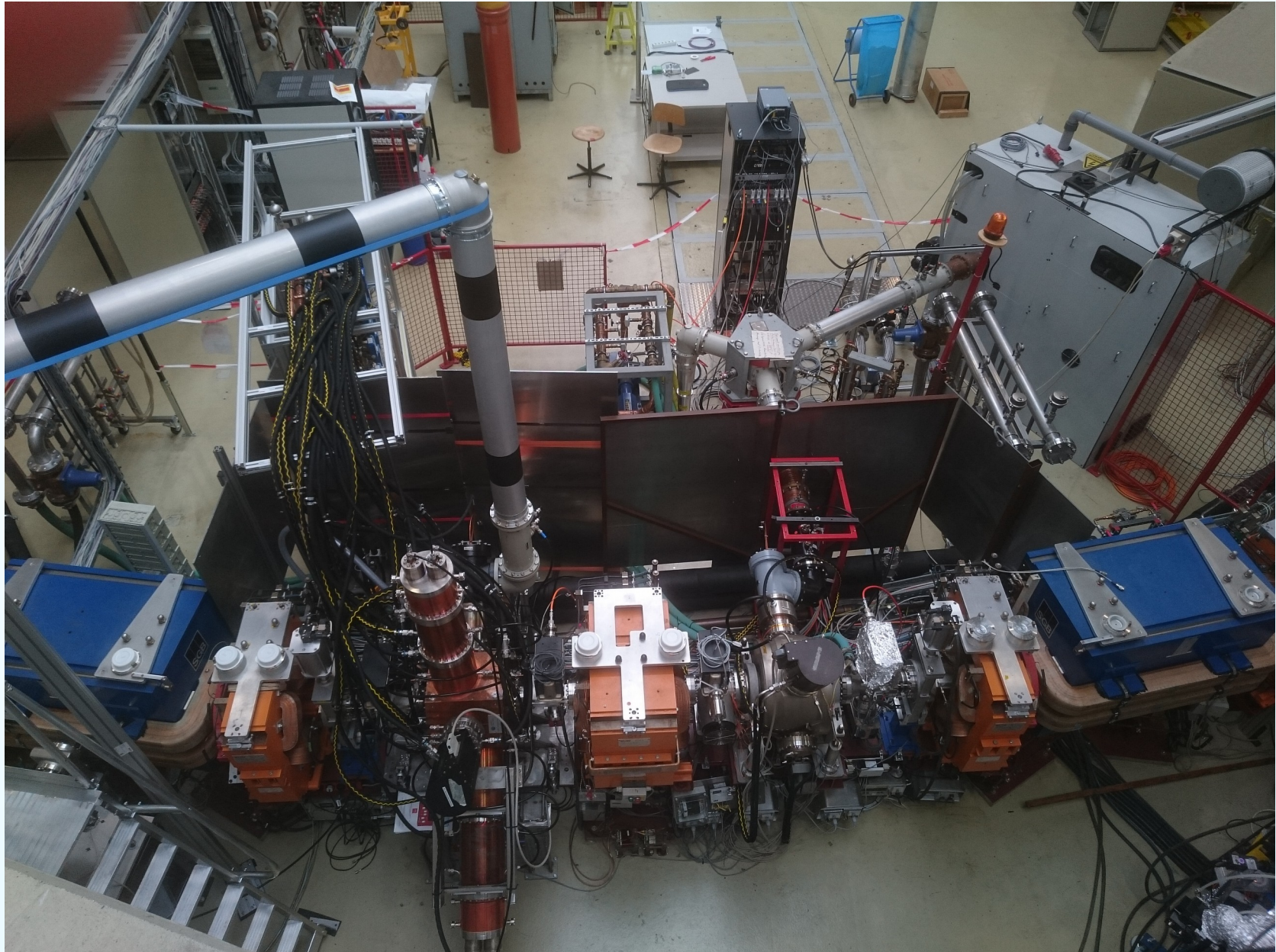
LINAC: Thales TH2100-D  
S-band klystron amplifier

Booster: Analog LLRF @ 499,8 MHz  
Solid state amplifier (Cryoelectra, 25 kW)  
DESY type 3-cell cavity

SR: Klystron amplifier YK-1265 (65 kW)  
DORIS type single-cell cavity









Work done on RF systems:

03/2021: LINAC S-Band Klystron replaced

03/2021: Defective module in Booster-SSA after power outage

08/2020: DELTA phase modulation scheme transferred to LLRF

02/2020: RF GUI: Cavity phasing changed

10/2019: SR: RF Interlock scheme changed due to insufficient damping of LLRF internal PIN Diode for rad. prot. purpose

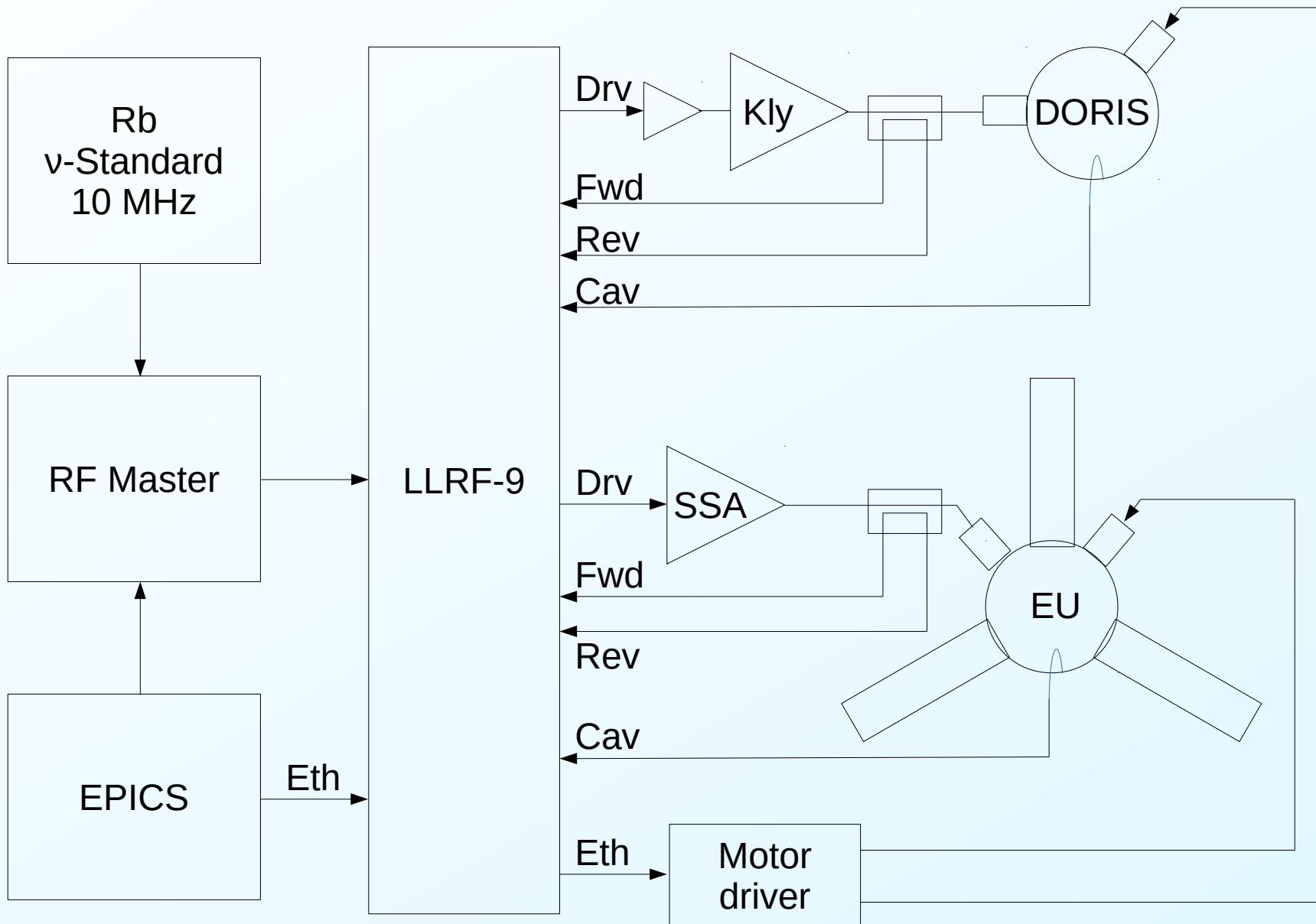
08/2019: Booster SSA: Broken PLC changed



# Storage Ring LLRF calibration



# Storage Ring RF







# LLRF calibration check with beam

Synchrotron frequency:  $f_s = f_{rev} \sqrt{\frac{\alpha h e U_C}{2\pi E} \sin(\phi_s)}$  ,  $\sin(\phi_s) = \sqrt{1 - \left(\frac{U_{rev}}{U_C}\right)^2}$

$$U_C = \sqrt{\left(\frac{2\pi E}{f_{rev} \alpha h e}\right)^2 f_s^4 + U_{rev}^2}$$

$k$

Reference parameters:

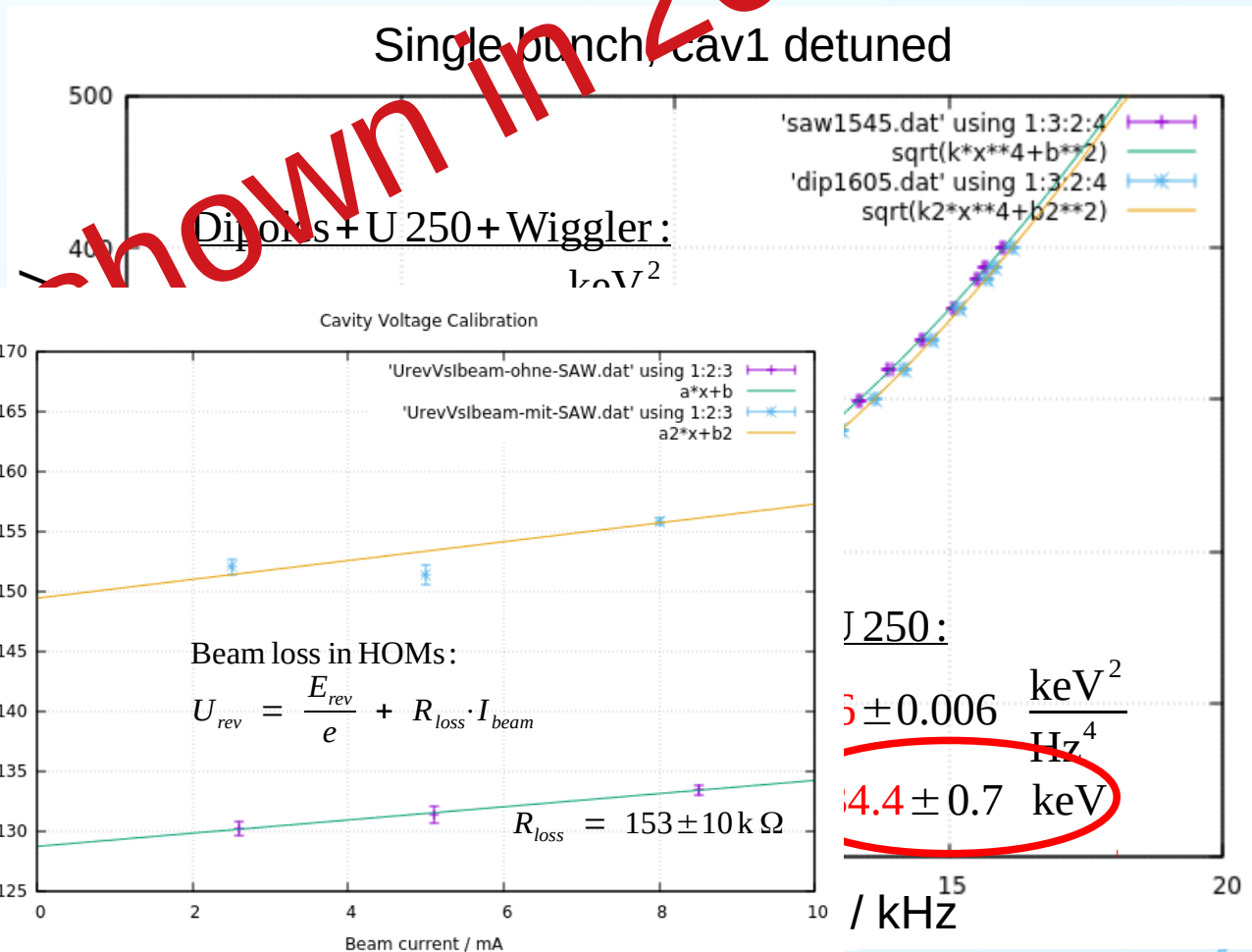
(E = 1492 MeV,  $\alpha = 0.0050$ )

$$\rightarrow k^{calc} = 2.076 \frac{\text{keV}^2}{\text{Hz}^4}$$

Calculated and from Simulation:

$$U_{rev}^{Dipoles + U_{250} + Wiggler} = 150 \text{ keV}$$

$$U_{rev}^{Dipoles + U_{250}} = 128 \text{ keV}$$



Already shown in 2019

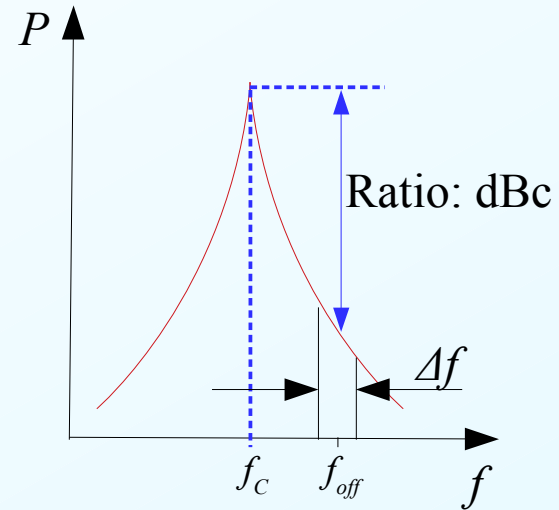
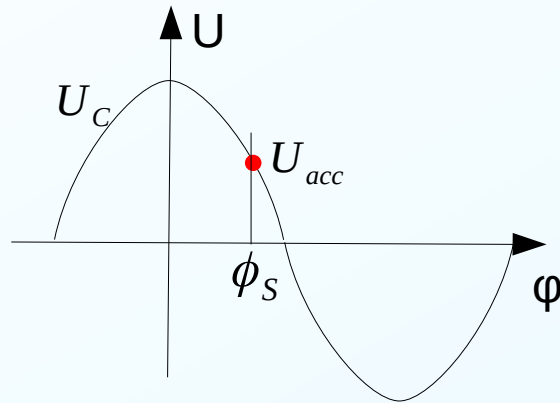


# Phase Noise





# Phase Noise

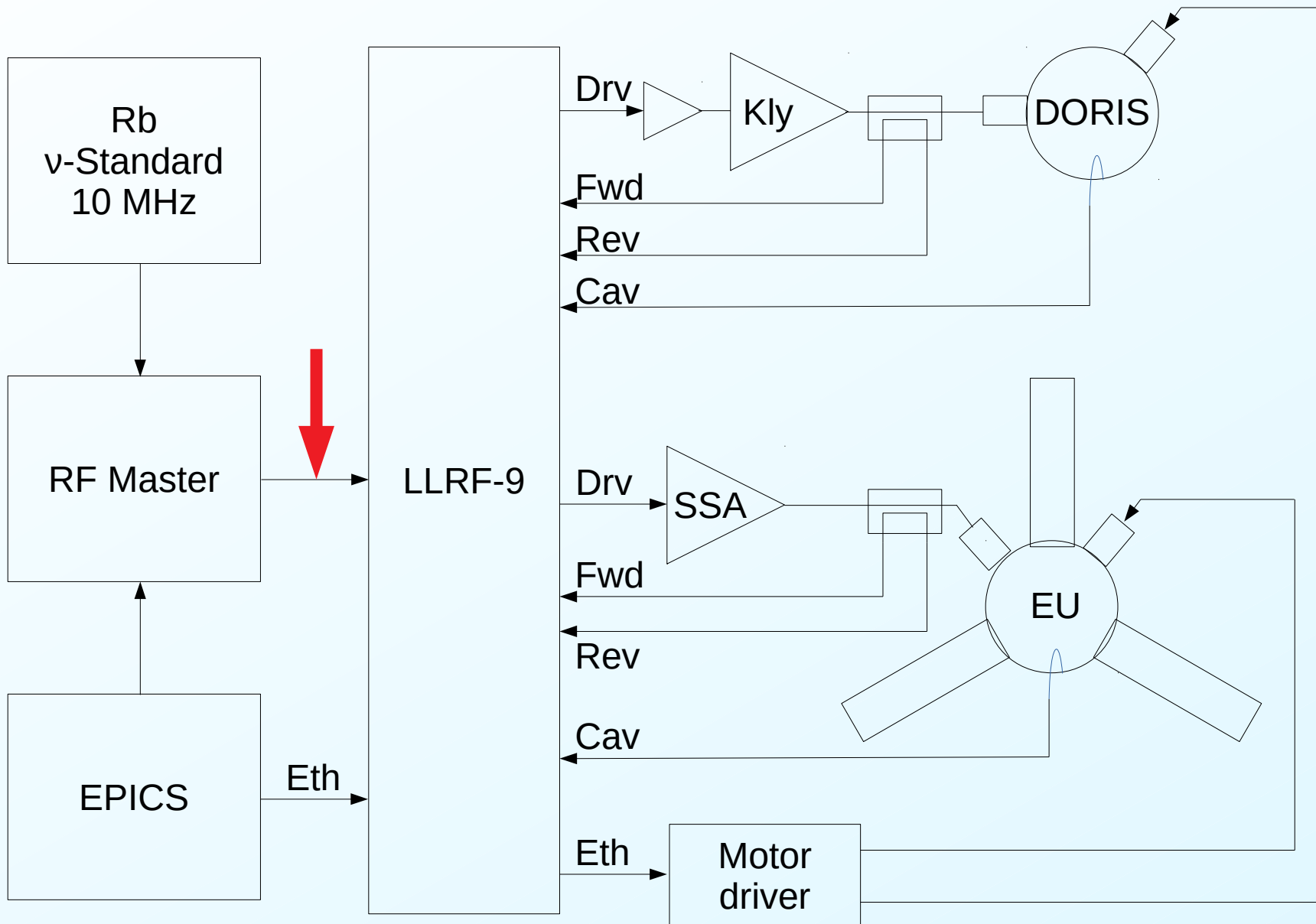


Noise power density:  $\mathcal{L}(f) = \frac{\text{Power density}}{\text{Carrier power}} \left[ \frac{\text{dBc}}{\text{Hz}} \right] = \frac{1}{2} S_{\phi}(f)$

Timing Jitter:  $\sigma_t = \frac{1}{2\pi f_c} \sqrt{\int_{f_1}^{f_2} S_{\phi}(f) df}$



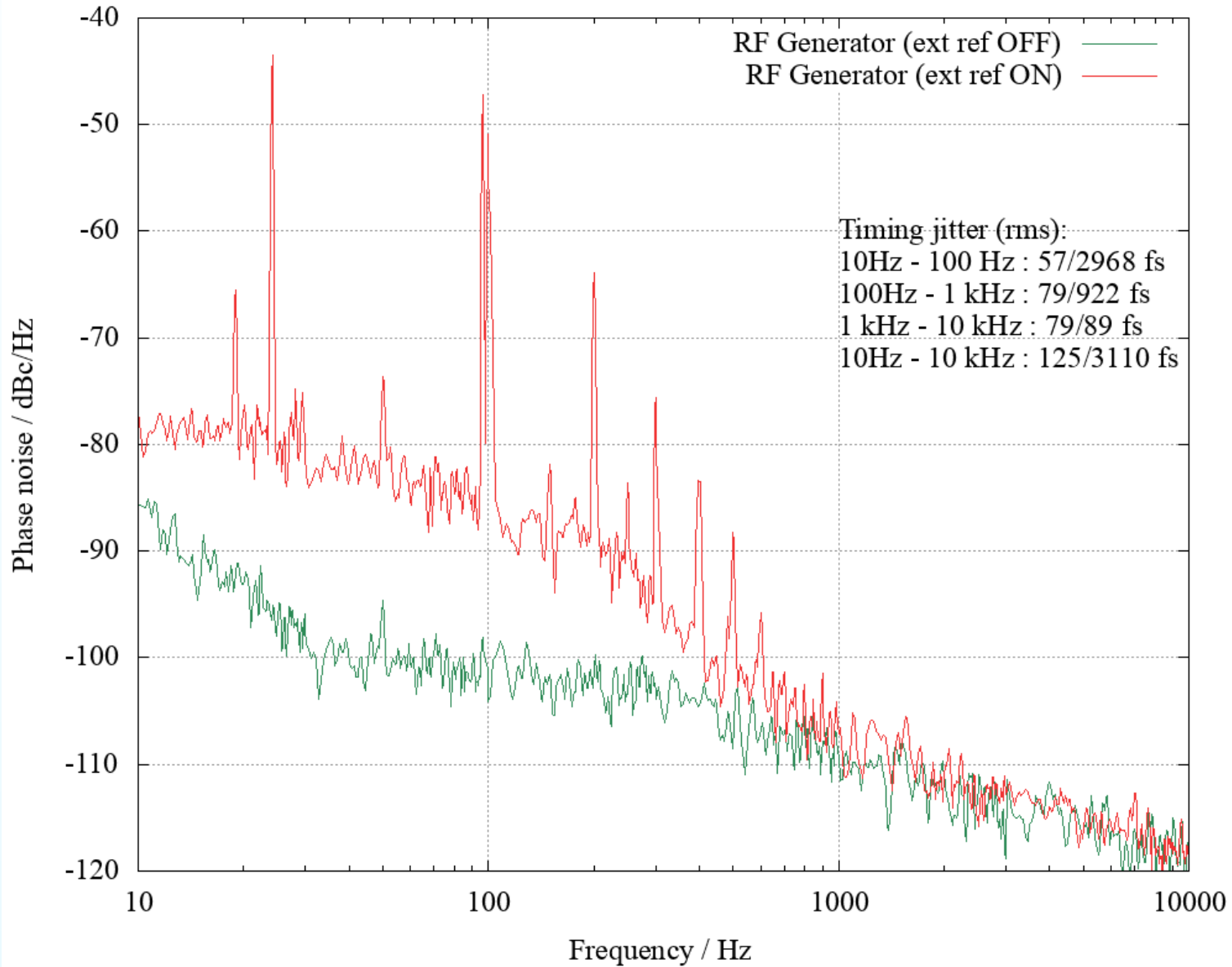
# Storage Ring RF





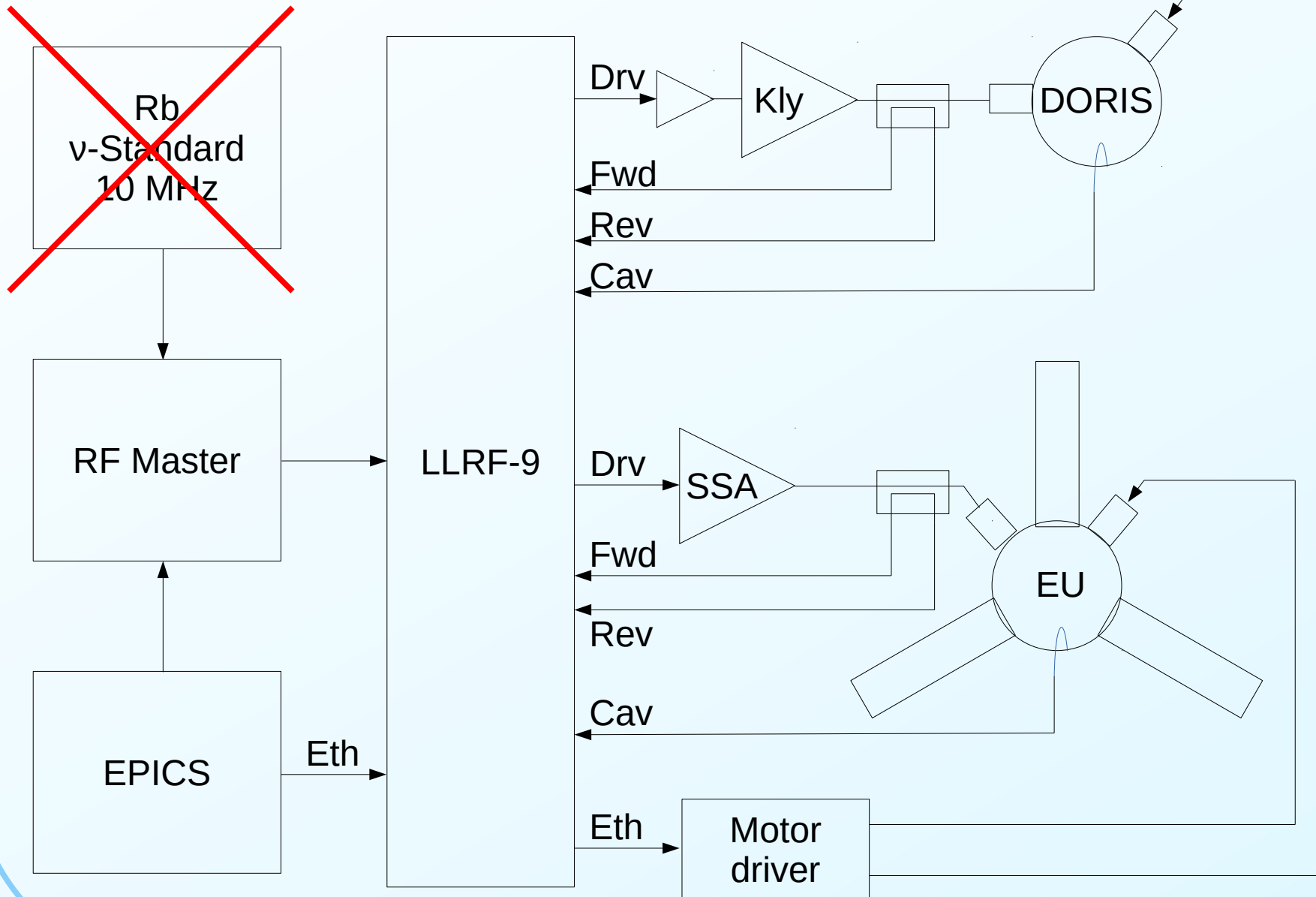
# Phase Noise

(Analyzer and Master generator)





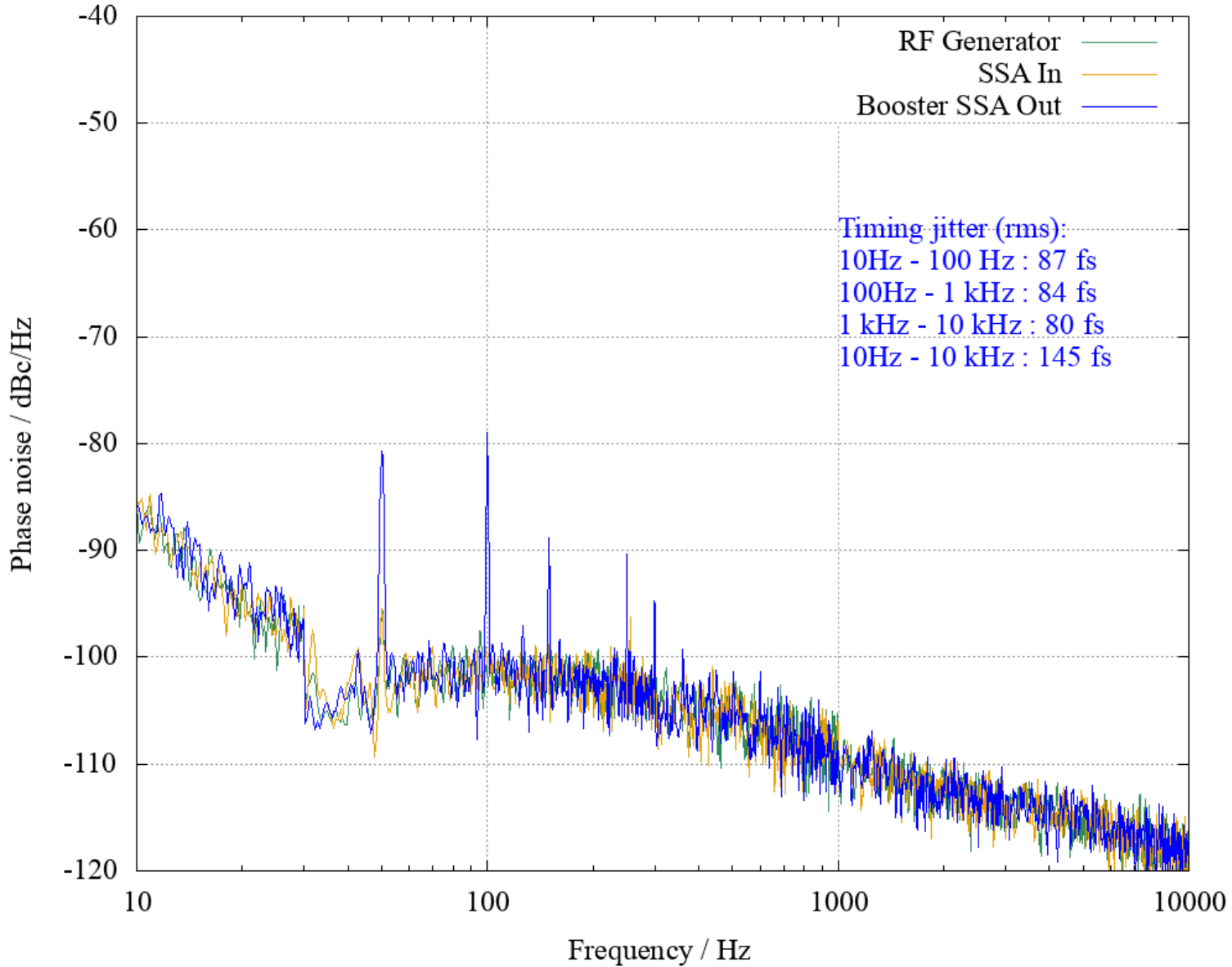
# Storage Ring RF





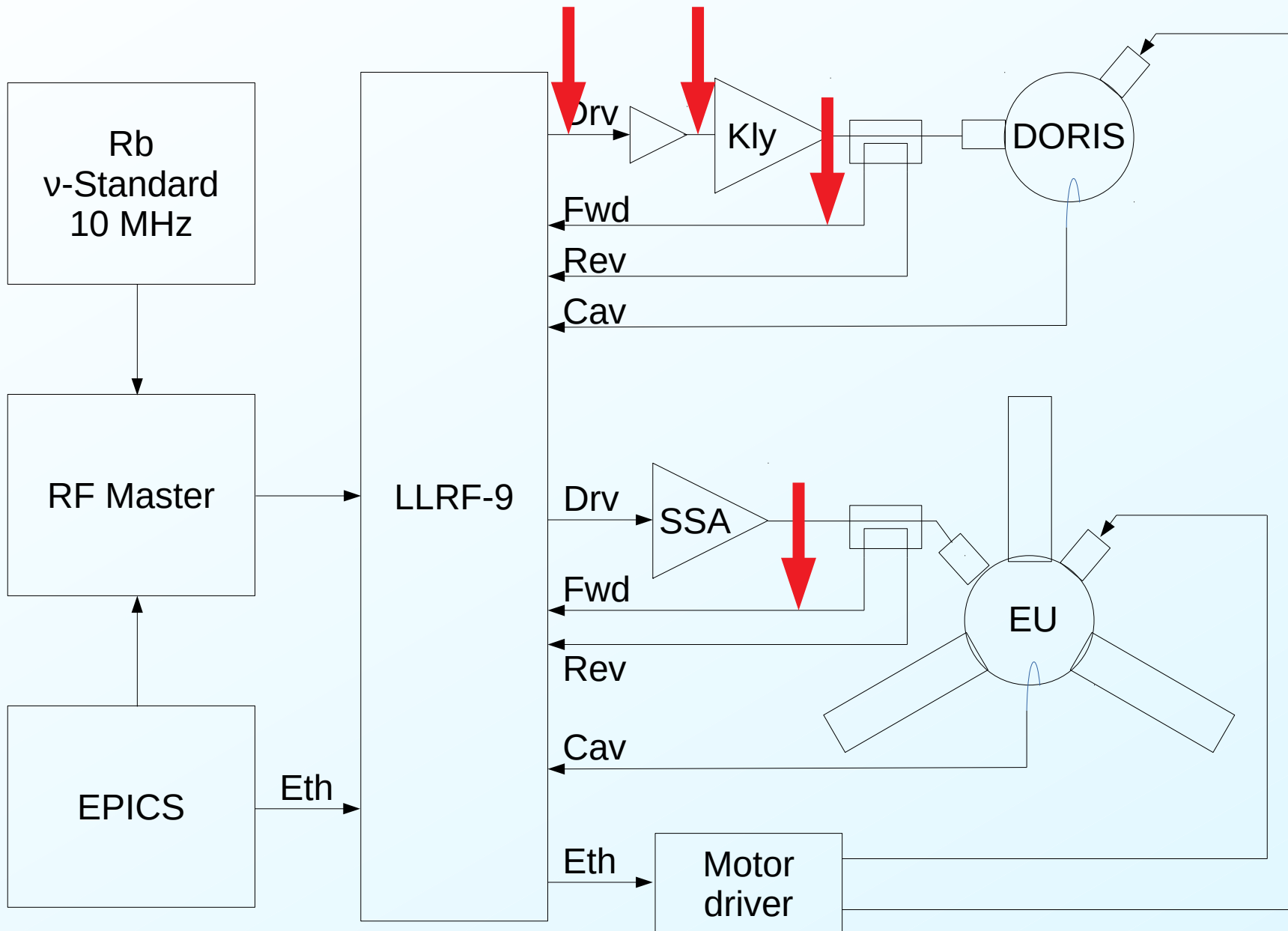


# Phase Noise (Booster)



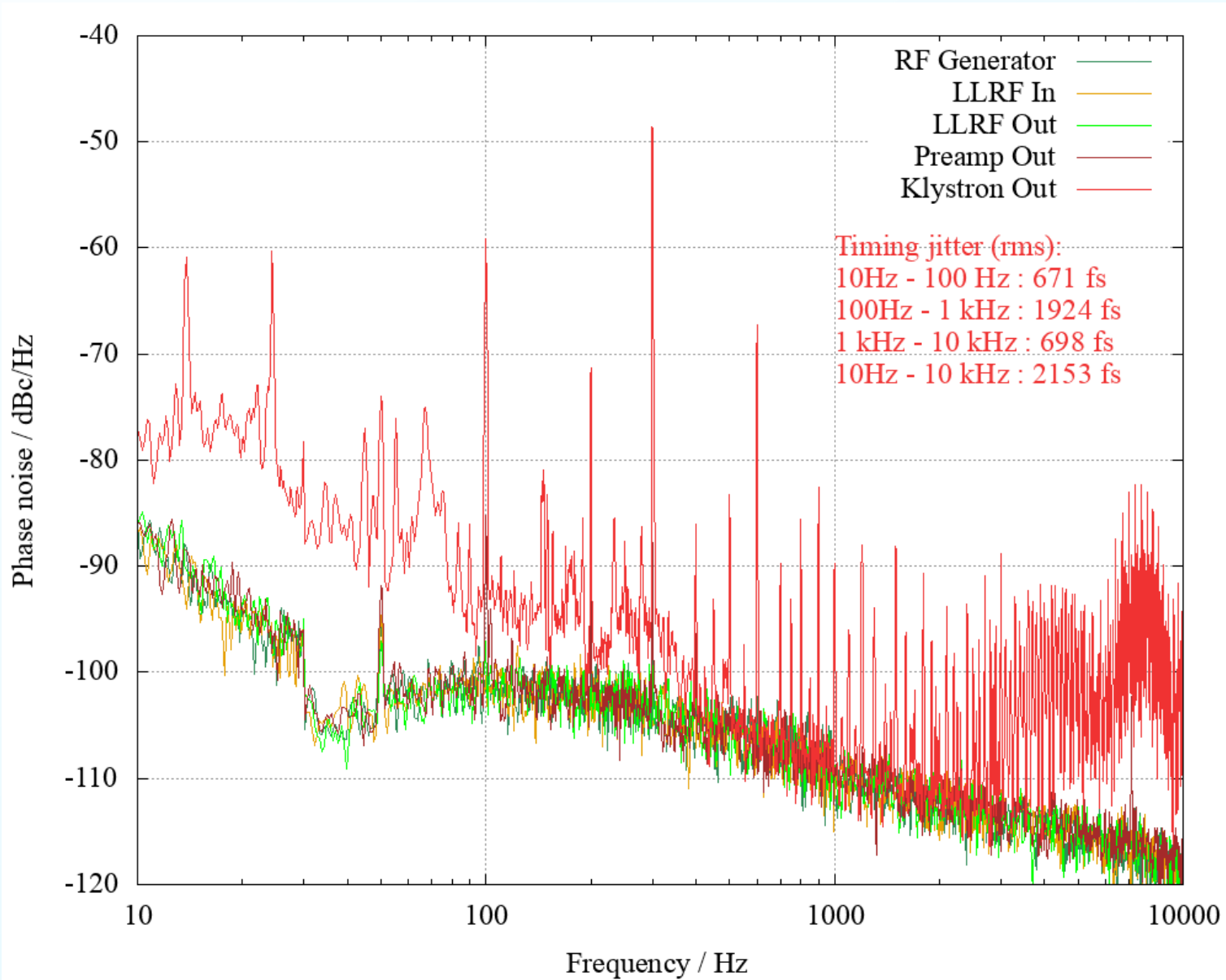


# Storage Ring RF





# Phase Noise (Storage Ring)





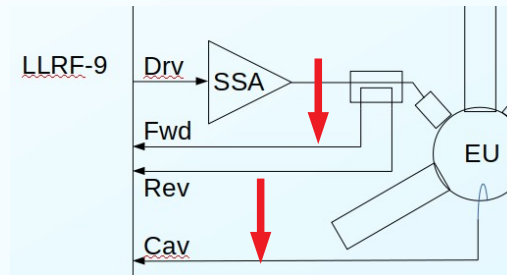
# Cavity Shunt Impedance





# Measure Shunt Impedance

## Method 1: Measure cavity voltage and fwd power



$$R_S = \frac{U_{cav}^2}{2 P_{fwd}}$$

$U_{NRVS} / V$	$U_{cav} / kV$	$P_{fwd} / kW$	$R_S / M\Omega$
1,937	307,0	11,91	3,96
2,259	358,0	16,25	3,94
2,517	398,9	20,25	3,93
1,614	255,8	8,25	3,97
1,291	204,6	5,27	3,97

$$R_S = 3.954 \pm 0.034 M\Omega$$



# Cavity Shunt Impedance

## Beam based measurements



# Cavity impedance without beam loading

$$Z_C(\omega_r + \Delta \omega) = \frac{R_S}{1 + iQ\left(\frac{\omega_r + \Delta \omega}{\omega_r} - \frac{\omega_r}{\omega_r + \Delta \omega}\right)}$$

$$\stackrel{\Delta \omega \ll \omega_r}{=} \frac{R_S}{1 + iQ\left(\frac{2\Delta \omega}{\omega_r}\right)}$$

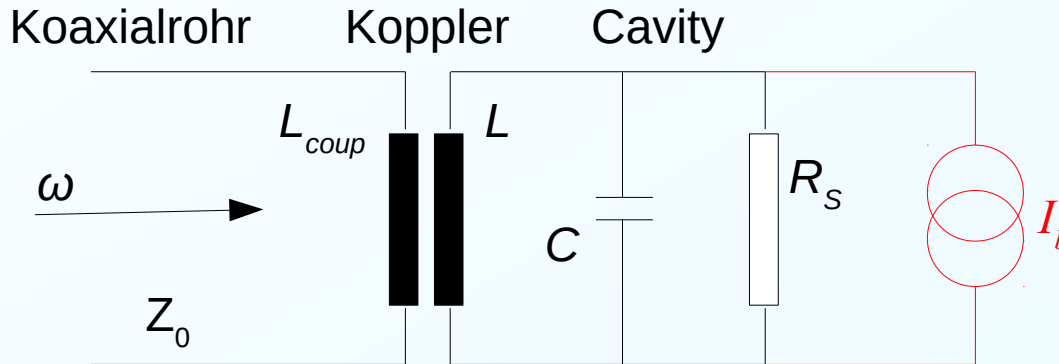
$$= \frac{R_S}{1 + Q^2\left(\frac{2\Delta \omega}{\omega_r}\right)^2} - i \frac{R_S Q\left(\frac{2\Delta \omega}{\omega_r}\right)}{1 + Q^2\left(\frac{2\Delta \omega}{\omega_r}\right)^2}$$

→ Cavity phase:

$$\tan(\phi_C(\omega_r + \Delta \omega)) = \frac{\Im(Z_C)}{\Re(Z_C)} = -Q \cdot \frac{2\Delta \omega}{\omega_r}$$



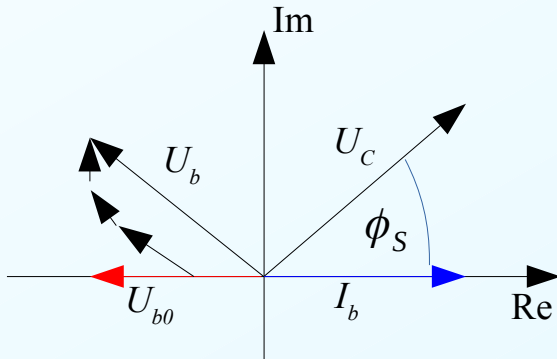
# Beam loaded cavity impedance



$$Z_C = \frac{R_s}{1 + iQ_0 \xi}$$

$$\beta = \frac{R_s}{Z_0} \cdot \frac{L_{coup}}{L}$$

$$Z = \frac{L_{coup}}{L} \left( Z_C \parallel \frac{\tilde{U}_C}{\tilde{I}_b} \right) = \frac{\beta Z_0}{R_s} \frac{1}{\frac{1}{Z_C} + \frac{\tilde{I}_b}{\tilde{U}_C}} = \frac{\beta Z_0}{1 + iQ_0 \xi + R_s \frac{\tilde{I}_b}{\tilde{U}_C}}$$



$$\tilde{U}_C \propto \tilde{I}_b e^{i\phi_s} \Rightarrow R_s \frac{\tilde{I}_b}{\tilde{U}_C} = R_s \frac{|\tilde{I}_b|}{|\tilde{U}_C|} \cdot e^{-i\phi_s}$$

$|\tilde{I}_b|$ : Leading fourier component  $2I_{DC}$  of the base frequency  $\omega$ ,  
Generated by an infinite row of short bunches with a temporal  
Distance of  $T_b = 2\pi/\omega$ .





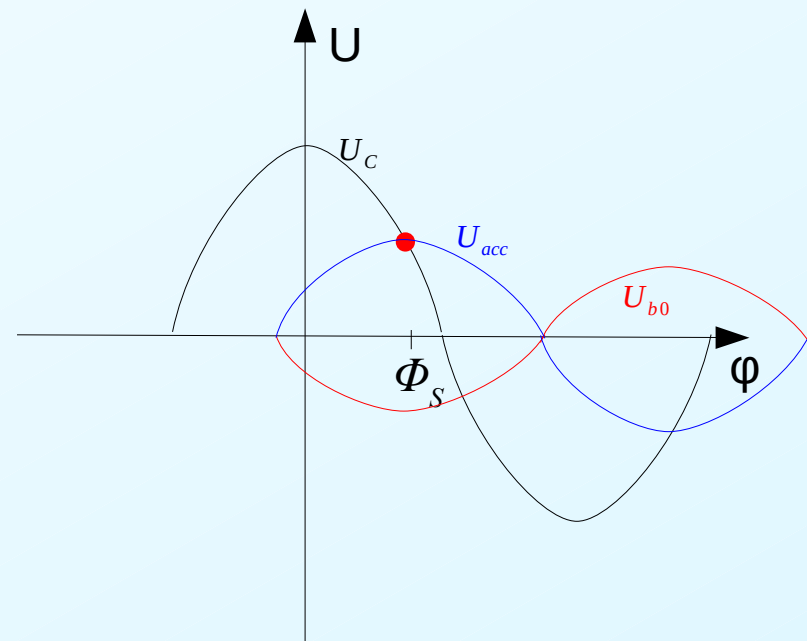
# Beam Loaded cavity impedance

$$Z = \frac{\beta Z_0}{1 + iQ_0 \xi + R_s \frac{\tilde{I}_b}{\tilde{U}_C}} = \frac{\beta Z_0}{1 + iQ_0 \xi + \frac{2 I_{DC} R_s}{U_C} e^{-i\phi_s}}$$

$$Z = \frac{\beta Z_0}{1 + \frac{2 I_{DC} R_s}{U_C} \cos(\phi_s) + i \left( Q_0 \xi - \frac{2 I_{DC} R_s}{U_C} \sin(\phi_s) \right)}$$

In case of match ( $Z = Z_0$ ):

1.  $\beta = 1 + \frac{2 I_{DC} R_s}{U_C} \cos(\phi_s)$
2.  $0 = Q_0 \xi - \frac{2 I_{DC} R_s}{U_C} \sin(\phi_s)$

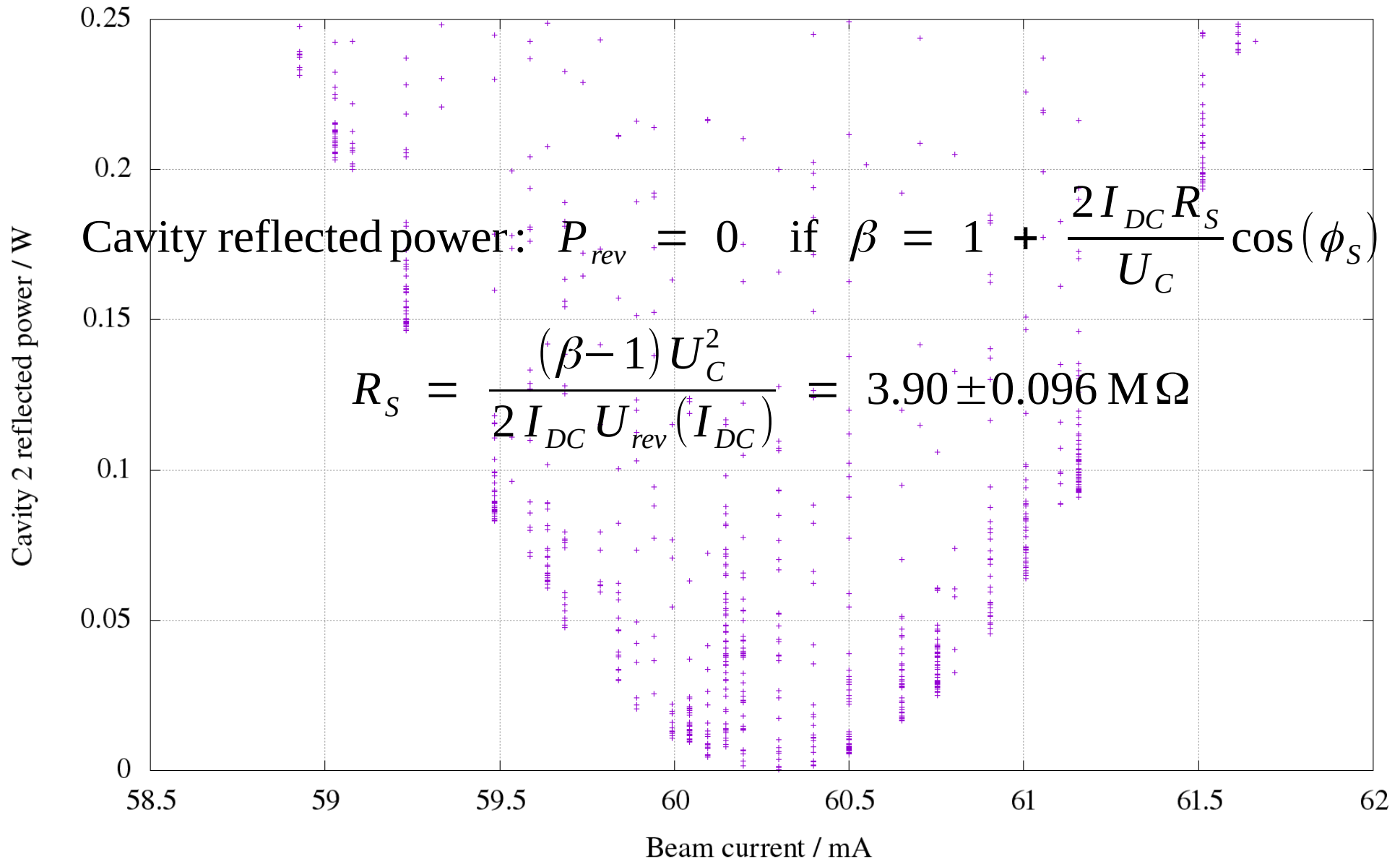




# Measure Shunt Impedance

## (Method 2: Use **real** part of Z)

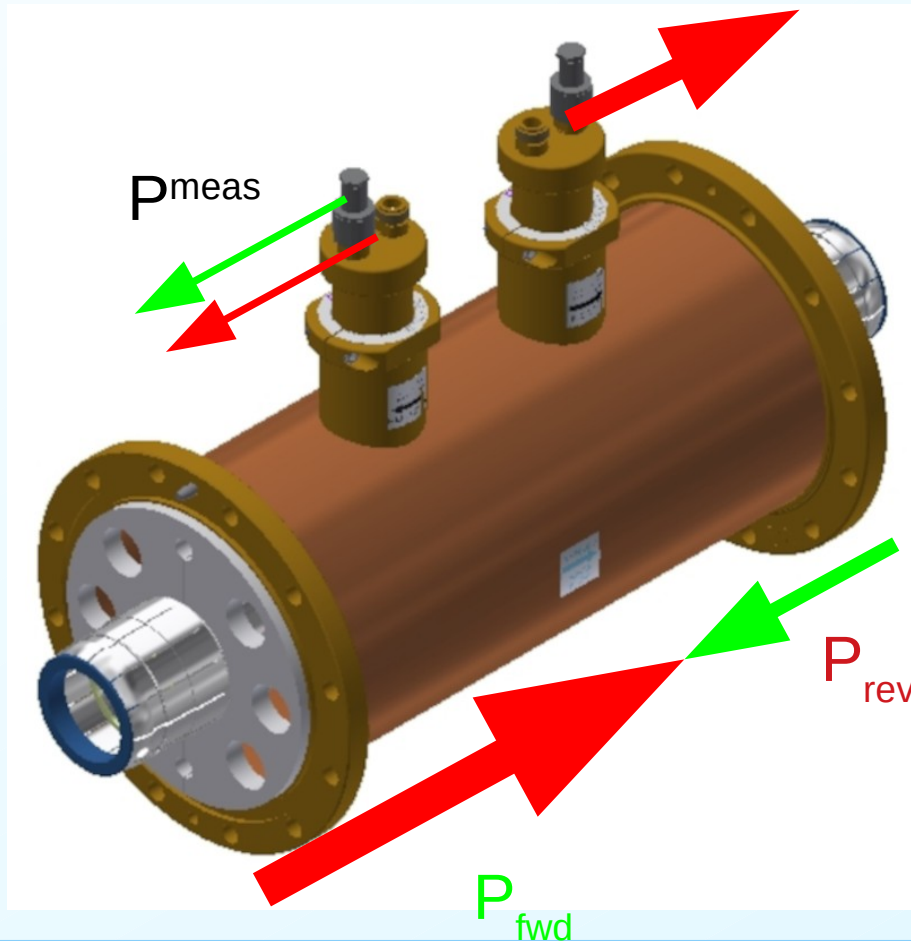
Cavity 2 reflected power at 387 kV cavity voltage (cav1 detuned)





The measurement of the minimum reflected power vs. beam current is **disputable**, because of the limited directivity (typ: -30dB, ours: < -42dB) of the directional coupler used to measure the reflected power:

A large forward power generates a false signal of the order of magnitude of the measured returned power in the reflected power line.





# Shunt Impedance

(Method 3: Use **imaginary** part of Z)

Imaginary part condition:  $0 = Q_0 \xi - \frac{2I_{DC}R_S}{U_C} \sin(\phi_S)$

with  $Q_0 \xi \stackrel{\Delta \omega \ll \omega_r}{=} (1+\beta)Q_L \cdot \frac{2\Delta\omega}{\omega_r} = -(1+\beta) \tan(\phi_C)$

Cavity detuning angle:  $\tan(\phi_C) = -\frac{2I_{DC}R_S}{(1+\beta)U_C} \sin(\phi_S)$

Beam current dependent inductance of cavity, **compensated by tuner loop.**



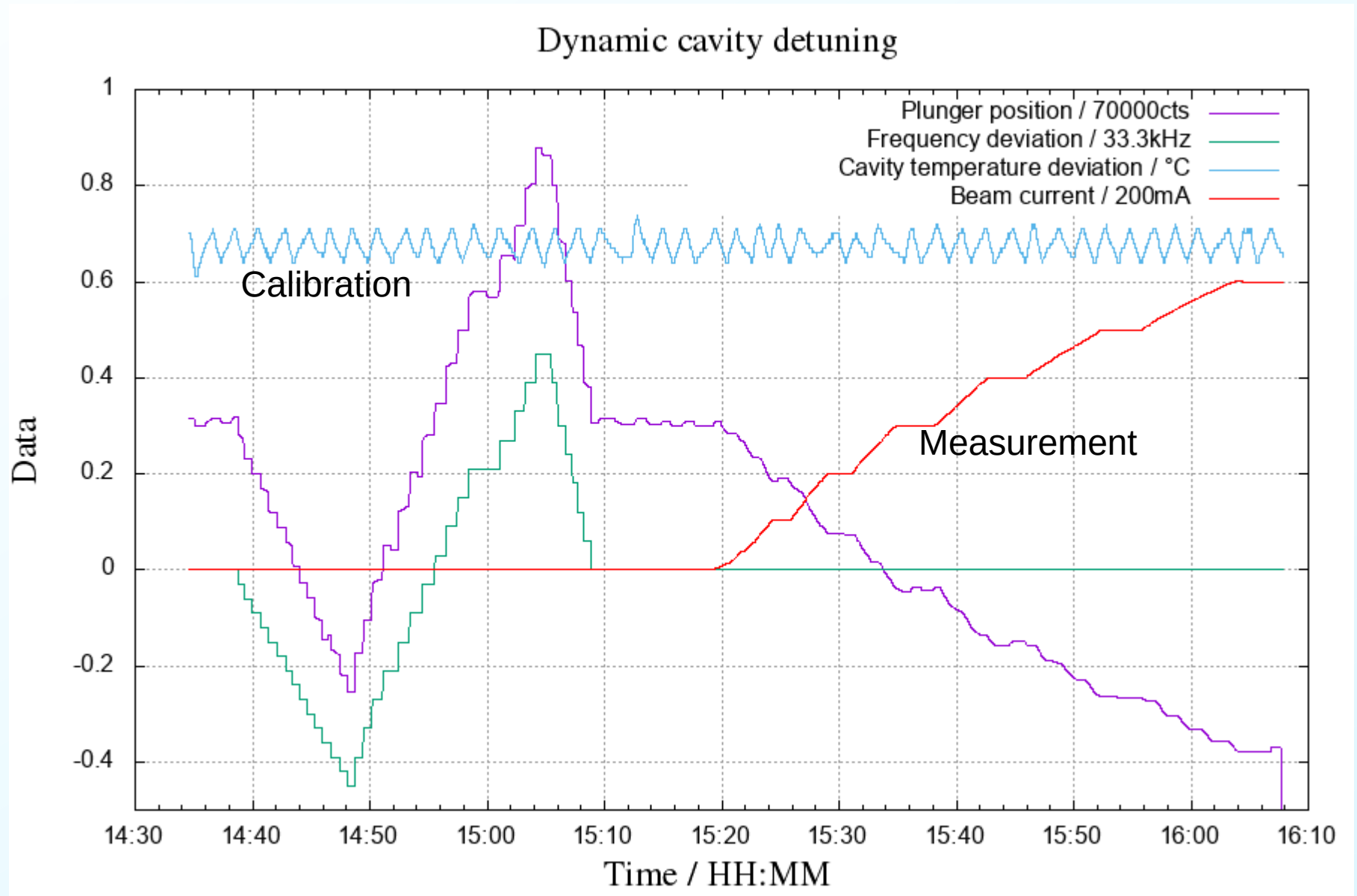


## Measurement procedure

1. Prerequisites:
  - Good measurement of loaded Quality factor  $Q_L$
  - Constant cavity temperature
2. Calibrate plunger stepper motor steps with frequency deviation  
From measured  $Q_L \rightarrow$  Relation btw. psm-steps and cavity phase  $\Phi_C$
3. Measure psm-Steps with increasing beam current
  - Slope gives measured cavity shunt impedance  $R_S$

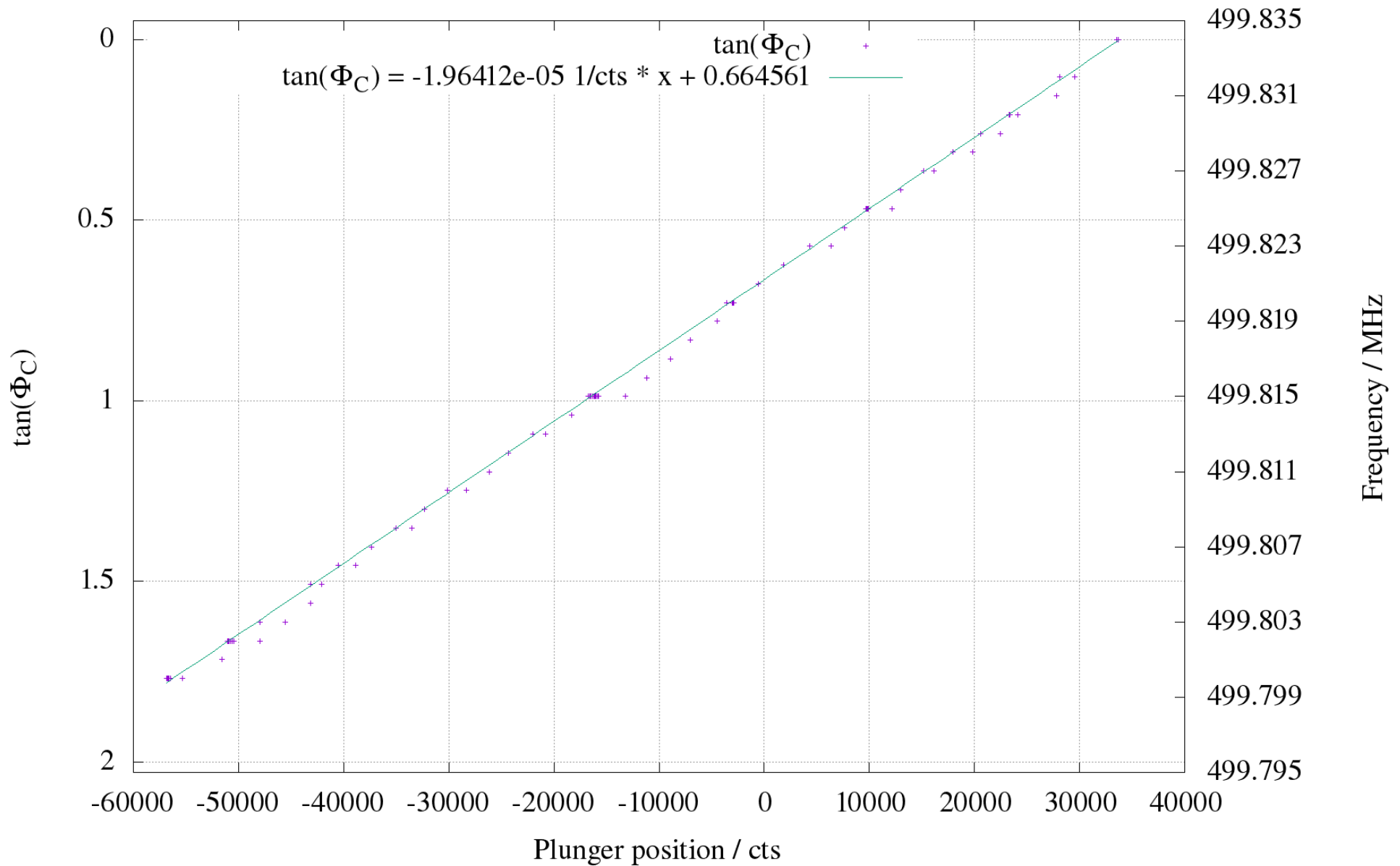


# Typical measurement



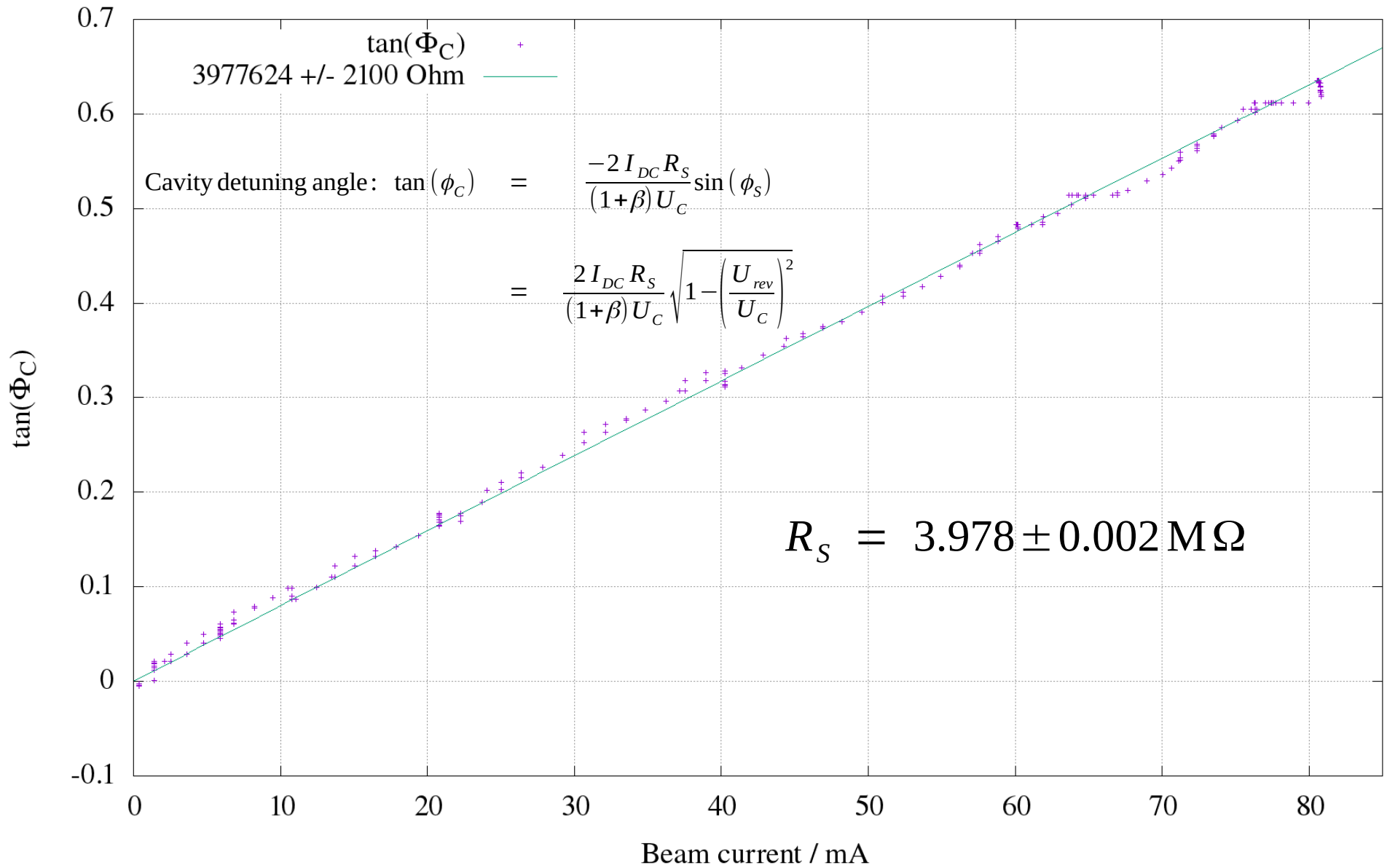


Frequency and  $\tan(\Phi_C)$  respectively vs. Plunger position,  $Q=13000$





### $\tan(\Phi_C)$ vs. Beam current





# Summary & Outlook

- The LLRF, SSA and EU cavity run with only minor hiccups for more than 2 years
- LLRF FPGA software changed by manufacturer to fulfill rad. prot. demands
- The LLRF voltage and power measurement calibration is good
- Phase noise is generated by the external reference (already uninstalled) and by the klystron (power supply?!)
- The shunt impedance  $R_s$  of the EU-type cavity was measured to be 3,97 M $\Omega$
  
- Next step: Install second LLRF in booster
- If funded: Replace storage ring klystron with SSA



# Acknowledgements

Vadim Kniss, Andreas Leinweber

DELTA Team

University electronics workshop

University mech. workshop

## Literature:

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Thank you  
for your attention